

9. (New) A method for detecting CDMA-coded signals  $\underline{d} = (\underline{d}^{(1)}, \dots, \underline{d}^{(K)})$ , where  $\underline{d}^{(k)} = (\underline{d}^{(k)}, \dots, \underline{d}^{(k)})$ ,  $k = 1, \dots, K$ , comprising the steps of:

determining a first detection solution  $\hat{\underline{d}}(1)$  of CDMA-coded signals  $\underline{d}$ ;

determining an  $(n+1)$ -th detection solution  $\hat{\underline{d}}(n+1)$  for  $n = 1, \dots, N$  as a function of the  $n$ -th detection solution  $\hat{\underline{d}}(n)$  by assigning

$$\hat{\underline{d}}(n+1) = f(\hat{\underline{d}}(n))$$

where iteration for  $n \rightarrow \infty$  converges toward multiuser solution  $\hat{\underline{d}}_{\text{MU}}$  corresponding to

$$f(\hat{\underline{d}}(n)) \xrightarrow{n \rightarrow \infty} \hat{\underline{d}}_{\text{MU}},$$

if a quality of approximation solution  $\hat{\underline{d}}(n+1)$  is not sufficient, assigning  $n \rightarrow n+1$  and continuing performing the step of determining the  $(n+1)$ -th detection solution ;

if the quality of approximate solution  $\hat{\underline{d}}(n+1)$  is sufficient, terminating the method and using  $\hat{\underline{d}}(n+1)$  as an estimate of data  $\underline{d}$  to be detected, wherein:

the function of the step of determining the  $(n+1)$ -th detection solution is given by

$$f(\hat{\underline{d}}) = \hat{\underline{d}} + \delta \cdot \underline{g}$$

with

$$\delta = \frac{\|\underline{g}\|^2}{\|A \cdot \underline{g}\|^2} \text{ and } \underline{g}^T = A^H \cdot (\underline{s}^T - A \cdot \hat{\underline{d}}^T)$$

where matrix A is given by

$$A = \begin{bmatrix} b^{(1)}_1 & 0 & \dots & \\ \vdots & \vdots & & \\ b^{(1)}_Q & 0 & & \vdots \\ \vdots & b^{(1)}_1 & & \\ b^{(1)}_{Q+W-1} & \vdots & & 0 \\ 0 & b^{(1)}_{Q+W-1} & & b^{(K)}_1 \\ \vdots & \vdots & & \vdots \\ 0 & 0 & \dots & b^{(K)}_{Q+W-1} \end{bmatrix}$$

with  $\underline{b}^{(k)} = \underline{c}^{(k)} * \underline{h}^{(k)}$ ,

where  $\underline{c}^{(k)}$  denotes the K different codes and  $\underline{h}^{(k)}$  denotes pulse responses of K different linear transmission channels.

10. (New) The method according to claim 9, further comprising the step of:  
converging toward a solution of a zero forcing block linear estimator for  $n \rightarrow \infty$ .
11. (New) The method according to claim 9, wherein:  
symbols  $d^{(k)}$  to be transmitted assume values of  $\pm 1$  or  $\pm i$ .
12. (New) The method according to claim 9, wherein:  
solution  $\hat{\underline{d}}^T(n) = A^H \cdot \underline{s}^T$  of a RAKE receiver is used as the 1-st detection  
solution for starting the iteration.
13. (New) The method according to claim 9, wherein:  
a first one of the detection solution for starting the iteration is set to zero.